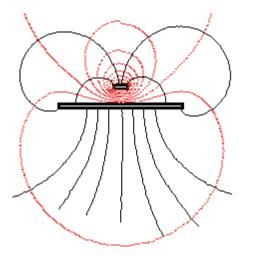
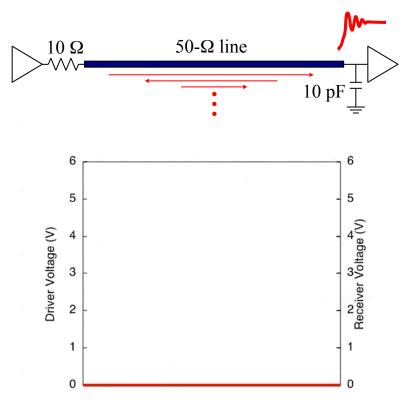
#### **Common Misconceptions about Inductance & Current Return Path**



Dr. Cheung-Wei Lam Apple Distinguished Engineer IEEE EMC Respected Speaker lam@alum.mit.edu



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#### Outline

- What are they?
  - L = Inductance

**I®** = Current Return Path

- Why do we care?
- Common Misconceptions
- How do we control them?
- How do we identify problems?
- Summary

### L: What is it?

- Various kinds: *loop*, *mutual*, external, internal, kinetic, self, *partial*, self partial, mutual partial, partial mutual, ...
- Definition of inductance for closed loops:

$$\begin{array}{c} I \\ 1 \\ \end{array} \\ \end{array} \\ 2 \\ L_1 = \frac{\Psi_1}{I_1} \\ M_{21} = \frac{\Psi_{21}}{I_1} \\ \end{array}$$

- External, internal, kinetic
- Self, *partial*, self partial, mutual partial, partial mutual

#### L: Recommended References

- Book:
  - Clayton Paul, "Introduction to Electromagnetic Compatibility"
- Paper:
  - Al Ruehli, "Inductance Calculations in a Complex Integrated Circuit Environment," IBM Journal of R&D, September 1972.
- Articles:
  - Bruce Archambeault, "Decoupling Capacitor Connection Inductance," IEEE EMCS Newsletter, Spring 2009
  - Bruce Archambeault, "Part II: Resistive vs. Inductive Return Current Paths," IEEE EMCS Newsletter, Fall 2008

#### L: Why do we care?

- Affects signal quality, crosstalk, EMI.
- Voltage Drop/Fluctuation

$$V_{L} = L \frac{dI_{L}}{dt} \qquad \stackrel{+ V_{L} -}{\underset{I_{L}}{\longrightarrow}}$$

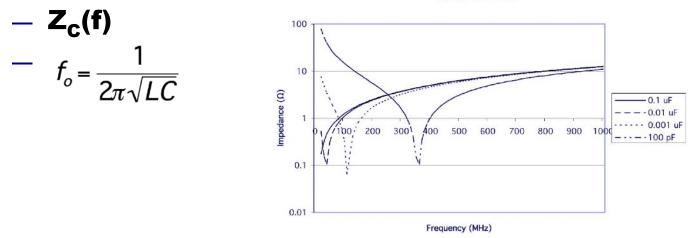
• Crosstalk & EMI

$$V_2 = M_{21} \frac{dI_1}{dt}$$
  $M_{21} = L_2$ 

#### L Why do we care?

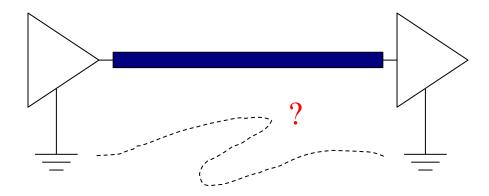
- Transmission Line Discontinuity  $\rightarrow$  Signal Ringing  $\bigwedge$
- Filtering & Decoupling

Zc for L = 2 nH



 $- Z_{c}(f)$ 

#### **I®:** What is it?

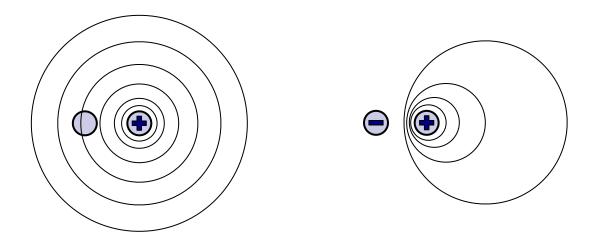


Is ground a zero-impedance equipotential surface?
 −V<sub>G</sub> = I<sub>G</sub>Z<sub>G</sub> = I<sub>G</sub>(R<sub>G</sub>+jωL<sub>G</sub>) ≠ 0

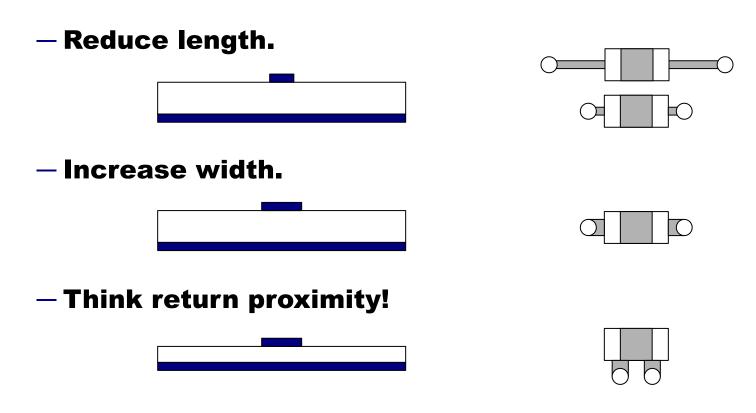
#### **I®: Why do we care?**

- Increases current loop area A
   EMI ↑
- Increases loop inductance L
  - Signal Quality  $\downarrow$
  - EMI ↑
- Increases mutual inductance M
  - Crosstalk ↑
  - EMI ↑
- Increases ground (return) inductance L<sub>G</sub> or M<sub>G</sub>
  - –EMI ↑

- Mistake loop L as sum of self inductances  $(L_{self})$ !?
- Overlook the importance of return proximity!?



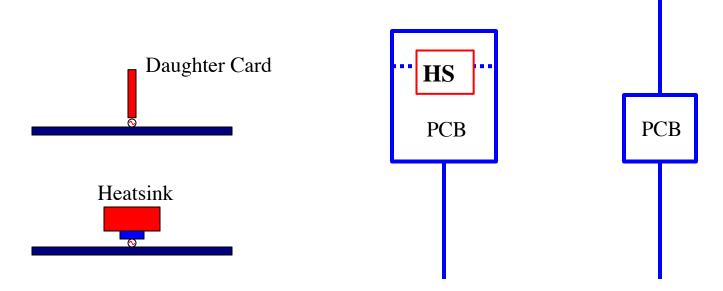
# L: Mounting Inductance

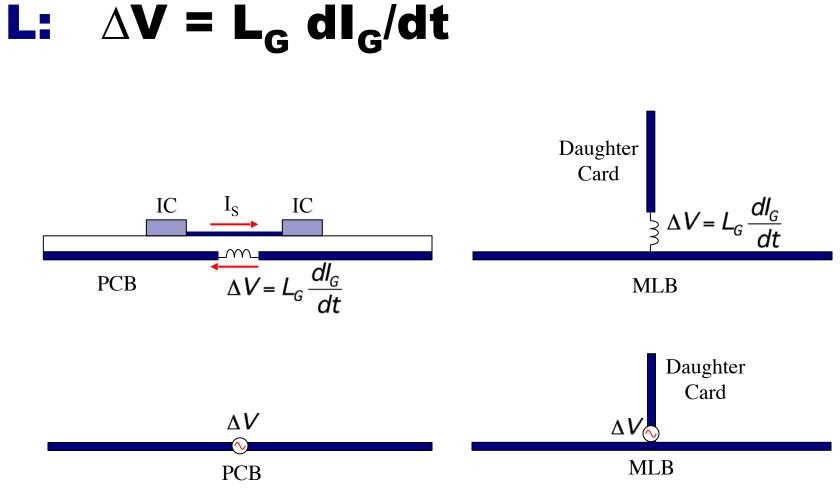


#### - Think loop inductance!!

• Ground Drop  $\propto$  Self Inductance  $(L_{Self})$ !?

#### - Ground Drop is a main source of CM radiation! - $V_{g} = I_{s} Z_{g} = I_{s} (R_{g} + j\omega L_{g})$

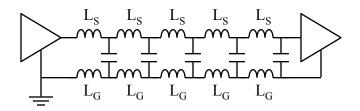




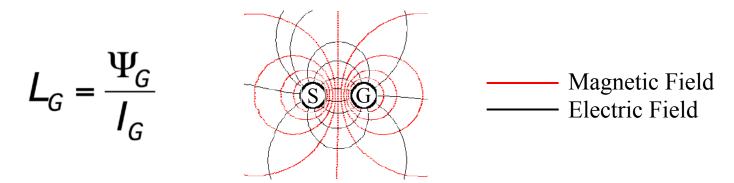
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#### L: Ground Inductance

• Transmission Line:  $L_T = L_S + L_G$ 

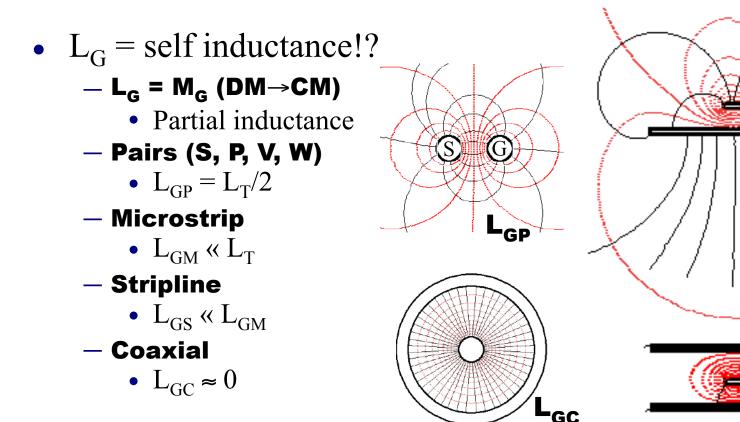


•  $\Psi_{\rm G}$  = magnetic flux around ground conductor



L: L<sub>G</sub> = ?

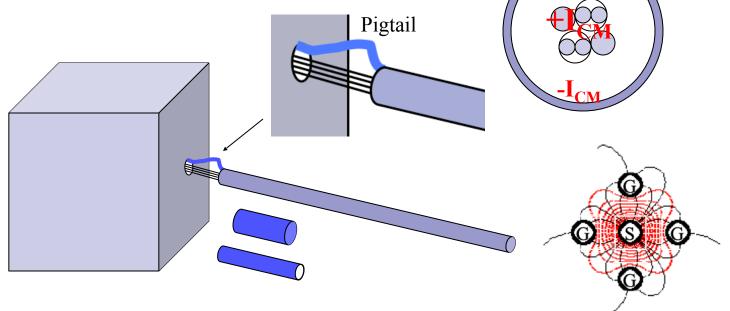
Magnetic FieldElectric Field



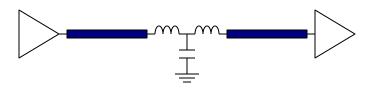
-GS

'GM

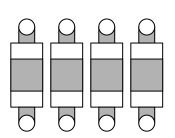
- Pigtail termination is bad because of its  $L_{Self}$ !? -  $M_{P}$  (inner-outer) (DM-CM)
  - Partial inductance



- Smaller is always better!?
   Excess capacitance
  - causes reflections!



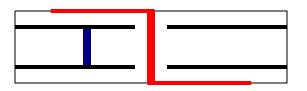
- Inductance parallels down like resistors!?
  - Don't forget M!
  - Spread out decoupling capacitors!
  - Alternate power/ground pins!

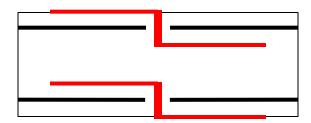


- Overlook mounting inductance vs. component inductance!?
  - Don't spend on expensive low-L filters unless layout has already been optimized.

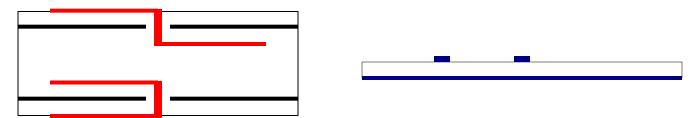
• Via Inductance = L<sub>Self</sub>!?

- Think loop inductance!

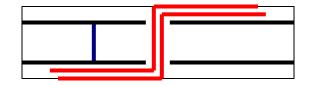




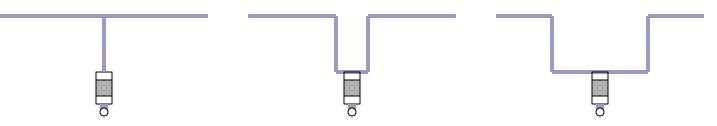
Overlook the dependence on current distribution!?
 – Current distribution affects inductance!

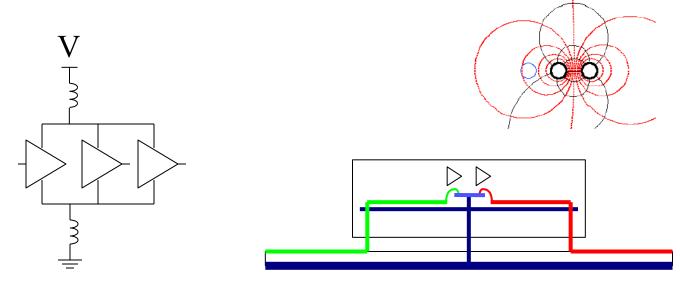


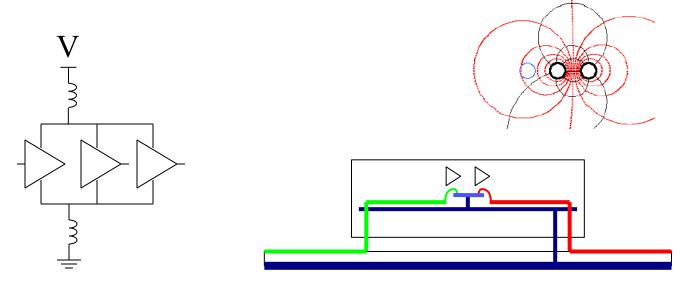
- Overlook the importance of return proximity!?
  - Think separation.
  - Think return proximity!

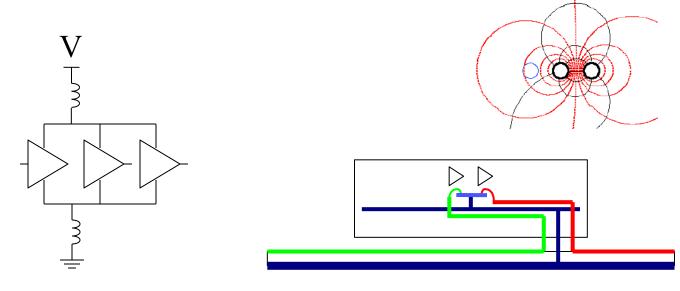


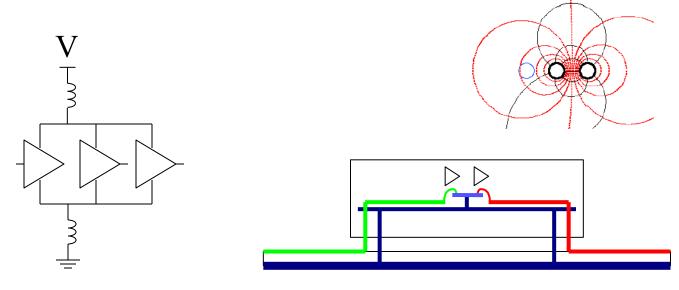
- L<sub>Self</sub> degrades capacitor performance!?
  - Think mutual inductance!









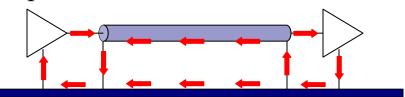


• Signal ground is a current source/sink!?

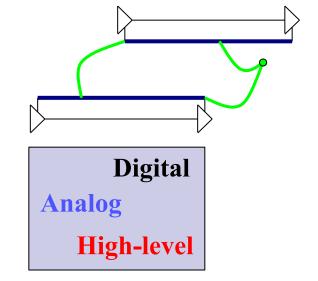


- Ground plane is a zero-impedance equipotential surface!?
  -V<sub>G</sub> = I<sub>G</sub>Z<sub>G</sub> = I<sub>G</sub>(R<sub>G</sub>+jωL<sub>G</sub>) ≠ 0
  -At kHz: R<sub>G</sub> » jωL<sub>G</sub>
  - IR drop causes common-impedance coupling.
  - At MHz/GHz:  $R_{G} \ll j_{\Omega}L_{G}$ 
    - I® affects A, L, M, SI, EMI.

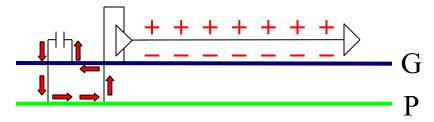
- Current takes the least resistance path!?
  - $\mathbf{Z}_{\mathbf{G}} = \mathbf{R}_{\mathbf{G}} + \mathbf{j} \mathbf{\omega} \mathbf{L}_{\mathbf{G}}$
  - Think R at  $f \leq kHz$ .
  - Think L at  $f \ge MHz!$



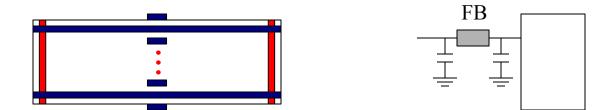
- Current returns along intended paths!?
  - IR drop  $\rightarrow$  common-Z coupling.
  - Current spreads out at  $f \le kHz$ .
  - Single-point grounding used for:
    - Low-level analog subsystems,
    - High-level noisy subsystems, e.g. motor drivers.



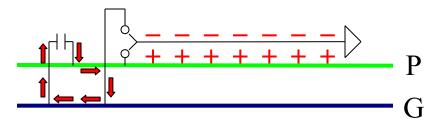
• Current returns through ground but not power!?



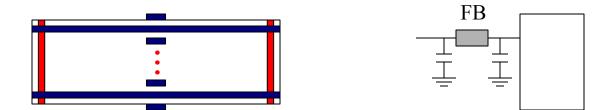
- Ground and power planes are interchangeable!?
  - Ground is connected to chassis, but not power.
  - Power isolation breaks the symmetry.



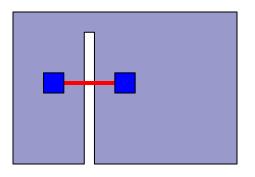
• Current returns through ground but not power!?

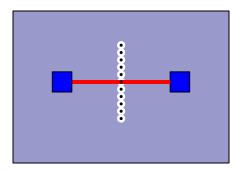


- Ground and power planes are interchangeable!?
  - Ground is connected to chassis, but not power.
  - Power isolation breaks the symmetry.



#### • Overlook horizontal return path!?

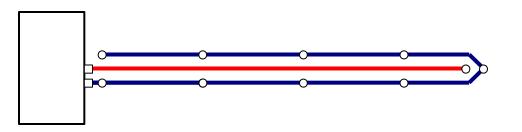


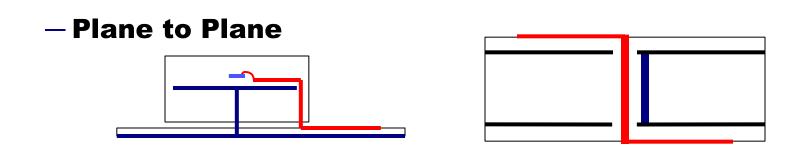


- Traces crossing plane cuts
  - Avoid ground plane cuts.
  - Route around plane cuts.
  - Use stitching capacitors.

- Overlapping via antipads
  - Stagger vias.
  - Space vias apart.

# Overlook vertical return path!? — Trace to Plane





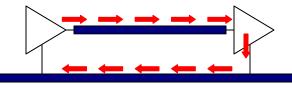
Overlook cross-board return paths!?
 – Avoid discontinuity.
 – Provide capacitors.



• Overlook off-board return paths!?

#### • Current flows in loops.

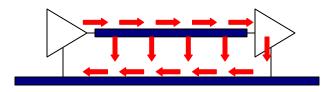
- Think of signal path and return path separately !?

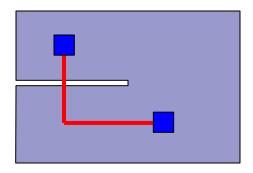


#### - Current flows in loops, but not this way.

#### - Current flows in pairs!

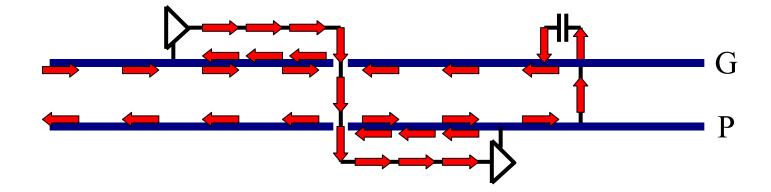
• Signal and return go hand-in-hand.





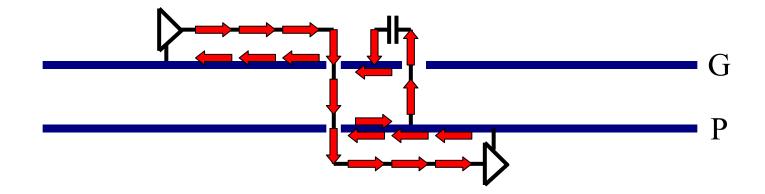
#### **R: Exercise**

• Trace out the current return path.



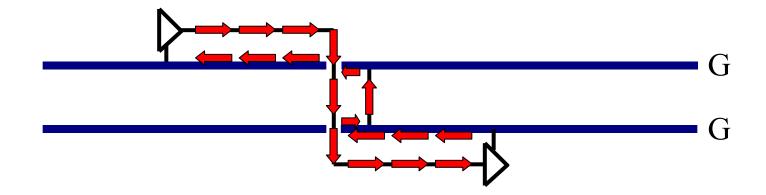
#### **R: Exercise**

• Trace out the current return path.

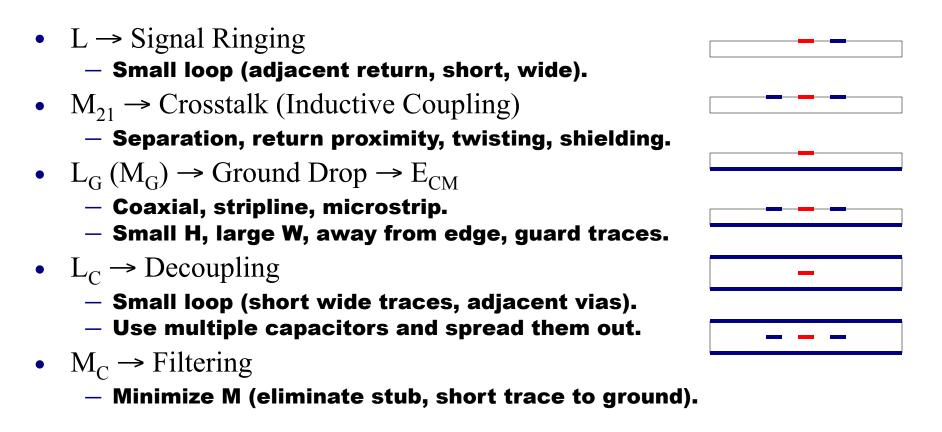


#### **R: Exercise**

• Trace out the current return path.



#### L: How do we control them?



#### **R**: How do we control them?

• At kHz:  $R_G \gg j\omega L_G$ 

#### Low-level analog or high-level noisy subsystems

- Single-point grounding prevents common-Z coupling.
- At MHz/GHz:  $R_G \ll j\omega L_G$

#### Horizontal return

- Use ground planes/grids instead of ground traces.
- Avoid traces crossing plane cuts.

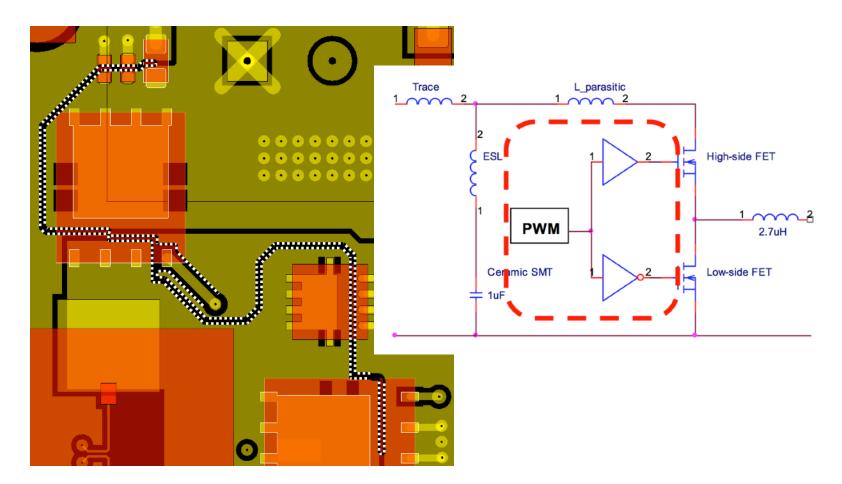
#### – Vertical return

- Provide adjacent return pins for noisy or susceptible pins.
- Provide adjacent vias, stitching capacitors as return bridges.
- Provide sufficient vias for guard traces.

#### L & IR: How do we identify problems?

Options	Pros	Cons
Fix when Fail	Less design time	Risks: time, cost,
Layout Review	Identify problems early	Labor intensive
Layout Checking Tool	Identify problems quickly Less labor intensive	Report 100's of violations Require expertise & time to identify critical violations
Automated & Customized Layout Checking Tool	Identify problems quickly No setup required Report critical violations	Require automation and customization development

#### **I**®: Tracing and Highlighting



## Summary

- L & I® affects signal quality, crosstalk and EMI.
- Inductance (L)
  - Forget self inductance.
  - Think loop, mutual, and partial inductance!
  - Think return proximity!
- Current Return Path (I®)
  - Low f: Current spreads out as  $R_{g} \gg j\omega L_{g}$ .
  - High f: Trace out I $^{\ensuremath{\mathbb{R}}}$  to identify discontinuities.